# Behavioral Approach to Systems Theory

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# About the course

#### lectures

- give enough background information for the exercises
- extras: optional presentations on special topics

#### exercises

- this is a core part of the course, not an optional extra
- links to exercises are showing in red in these slides

#### mini-projects

- to be discussed individually
- compulsory for those who need evaluation



## Introduction: the need

## Basics: notation and conventions

Data-driven interpolation and approximation



## Introduction: the need

## Basics: notation and conventions

Data-driven interpolation and approximation

The classical approach views system as input-output map

input 
$$\longrightarrow$$
 system  $\longrightarrow$  output

the system is a signal processor

accepts input and produces output signal

intuition: the input causes the output

The input-output map view of the system is deficient: it ignores the initial condition

example: mass driven by external force

- $\blacktriangleright \text{ input } \leftrightarrow \text{ force}$
- output  $\leftrightarrow$  position
- ► ??? ↔ position and velocity at start (initial condition)

input-output maps assume zero initial condition

how to account for nonzero initial condition?

Taking into account the initial condition leads to the state-space approach



paradigm shift from "classical" to "modern"

classical: scalar transfer function

modern: multivariable state-space

The modern state-space paradigm brought new theory, problems, and methods

## state-space theory

- manifests the "finite memory" structure of the system
- brought the concepts of controllability and observability
- deals seamlessly with time-varying and MIMO systems

## new problems / solution methods

- linear quadratic optimal control (LQ control)
- optimal state estimation (the Kalman filter)
- balanced model reduction

## amenable for numerical computations

A case in point: optimal filtering (signal from noise separation)

# Wiener filter (1942)

- transfer functions approach
- assumes stationarity
- no practical real-time method



# Kalman filter (1960)

- state-space approach
- non-stationary processes
- recursive real-time solution



There are other awkward things with the input/output thinking

modeling from first principles leads to relations

the input/output partitioning is not unique

interconnection of systems is variables sharing

# First principles modeling leads to relations

natural phenomena rarely operate as signal processors

the variables of interest satisfy relations, not functions

example: planetary orbits





# More basic example: Ohmic resistor voltage and current satisfy relation

to-be-modeled variables: voltage V and current I

Ohm's law:

- V = RI, with R the resistance
- I = CV, with C := 1/R the conductance

#### Q: how to fit the limit cases

- open circuit  $R = \infty$ , C = 0
- short circuit  $R = 0, C = \infty$

neatly in a unified framework?

A: V, I satisfy (linear) relation

# The behavioral approach was put forward by Jan C. Willems in the 1980's

3-part, 70-page, 1986–1987 Automatica paper:

Part I. Finite dimensional linear time invariant systems Part II. Exact modelling Part III. Approximate modelling

#### From Time Series to Linear System— Part I. Finite Dimensional Linear Time Invariant Systems\*

JAN C. WILLEMS†

Dynamical systems are defined in terms of their behaviour, and input/output systems appear as particular representations. Finite dimensional linear time invariant systems are characterized by the fact that their behaviour is a linear shift invariant complete (equivalently closed) subspace of  $(\mathbb{R}^{n_1^2} \circ (\mathbb{R}^{n_2^2+})$ .



Jan C. Willems (1939-2013)

# Critical revision of the input/output thinking

simple idea: the system is set of trajectories

variables not partitioned into inputs and outputs
 the system is separated from its representations

the input/output approach is a special case

relevant for the emerging data-driven paradigm

# The behavior is all that matters

"The operations allowed to bring model equations in a more convenient form are exactly those that do not change the behavior. Dynamic modeling and system identification aim at coming up with a specification of the behavior. Control comes down to restricting the behavior."

J. C. Willems, "The behavioral approach to open and interconnected systems: Modeling by tearing, zooming, and linking," Control Systems Magazine, vol. 27, pp. 46–99, 2007.

Analogy with solution of systems of equations

Q: what operations are allowed?

A: the ones that don't change the solution set (for linear systems, the "elementary operations")

the solution set is all that matters

Classical definition of linear system  $S: u \mapsto y$  is linear  $\iff S$  is linear function

for all u, v and  $\alpha, \beta \in \mathbb{R}$ ,  $S : \alpha u + \beta v \mapsto \alpha S(u) + \beta S(v)$ 



# The classical definition is deficient

# (silently) assumes

- zero initial condition
- controllability

#### doesn't apply to autonomous systems

relaxing the assumptions requires state-space

Behavioral definition of linear system  $\mathscr{B}$  is linear  $\iff \mathscr{B}$  is subspace

for all 
$$\textit{w},\textit{v} \in \mathscr{B}$$
 and  $\pmb{lpha},\pmb{eta} \in \mathbb{R}$ 

 $\alpha w + \beta v \in \mathscr{B}$ 

#### fixes the issues with

- nonzero initial condition
- autonomous systems
- controllable systems



# Separating problems from solution methods

## different representations ~~> different methods

- with different properties (efficiency, robustness, ...)
- their common feature is that they solve the same problem

## clarifies links among methods

leads to new methods

# Summary: behavioral approach

## detach the system from its representations

- define properties and problems in terms of the behavior
- lead to new, more general, definitions and problems
- avoid inconsistencies of the classical approach

#### separate problem from solution methods

- different representations lead to different methods
- show links among different methods
- lead to new solutions

## naturally suited for the "data-driven paradigm"

# Paradigms shifts

- 1940–1960 classical SISO transfer function
- 1960–1980 modern MIMO state-space
- 1980–2000 behavioral the system as a set
- 2000-now data-driven using directly the data



## Introduction: the need

## Basics: notation and conventions

Data-driven interpolation and approximation

# $(\mathbb{R}^q)^{\mathscr{T}}$ is the space of signals $w: \mathscr{T} \to \mathbb{R}^q$

## $\mathscr{T}$ — time axis

# $(\mathbb{R}^q)^{\mathscr{T}}$ — real-valued *q*-variate signals

## examples:

$$\mathbf{W} \in (\mathbb{R}^2)^{\mathbb{N}} \quad \leftrightarrow \quad \mathbf{W} = \left( \begin{bmatrix} w_1(1) \\ w_2(1) \end{bmatrix}, \dots, \begin{bmatrix} w_1(t) \\ w_2(t) \end{bmatrix}, \dots \right)$$
$$\mathbf{W} \in (\mathbb{R}^2)^T \quad \leftrightarrow \quad \mathbf{W} = \left( \begin{bmatrix} w_1(1) \\ w_2(1) \end{bmatrix}, \dots, \begin{bmatrix} w_1(T) \\ w_2(T) \end{bmatrix} \right)$$

It's a mistake to say "the signal w(t)"

let  $w \in (\mathbb{R}^q)^{\mathbb{N}}$  and  $t \in \mathbb{N}$ 

then,  $w(t) \in \mathbb{R}^q$  is the *value* of *w* at time *t* 

w(t) is not signal (in  $(\mathbb{R}^q)^{\mathbb{N}}$ ), but vector (in  $\mathbb{R}^q$ )

 $w(\cdot)$  — specifies explicitly the time dependence of w

# Use short, unambiguous, consistent notation

*"w* = *v*" means

"
$$w(t) = v(t)$$
, for all  $t \in \mathscr{T}$ "

shift operator  $\sigma$ 

$$(\sigma w)(t) := w(t+1)$$
, for all  $t \in \mathscr{T}$ 

# For example

### $\ell$ -th order vector difference equation

$$\begin{aligned} R_0 w + R_1 \sigma w + \cdots + R_\ell \sigma^\ell w &= 0 \\ & \\ & \\ R_0 w(t) + R_1 w(t+1) + \cdots + R_\ell w(t+\ell) &= 0, \text{ for all } t \in \mathbb{N} \end{aligned}$$

first order state equation

Compact notation for difference equation

$$R_0 w + R_1 \sigma w + \dots + R_\ell \sigma^\ell w = 0$$

$$R(\sigma) w = 0$$

polynomial operator $R(\sigma)=R_0+R_1\sigma+\dots+R_\ell\sigma^\ell$ 

kernel of polynomial operator ker $R(\sigma) := \{ w \mid R(\sigma)w = 0 \}$  We identify a dynamical system with its behavior, *i.e.*, the set of trajectories

real-valued system  $\mathscr{B}$  with q variables and time-axis  $\mathscr{T}$  is a subset of  $(\mathbb{R}^q)^{\mathscr{T}}$ 

in particular, we use set theoretic notation

 $w \in \mathscr{B} \iff w$  is a trajectory of  $\mathscr{B}$  $\iff \mathscr{B}$  is an exact model of w ... and specify  $\mathscr{B}$  by representations representation of the system  $\mathscr{B} \subseteq (\mathbb{R}^q)^{\mathscr{T}}$  $\mathscr{B} = \{ w \in (\mathbb{R}^q)^{\mathscr{T}} \mid \text{"constraints on } w" \}$ 

#### for example

kernel (KER) representation

$$\mathscr{B} = \ker R(\sigma) := \left\{ w \mid R_0 w + R_1 \sigma w + \dots + R_\ell \sigma^\ell w = 0 \right\}$$

input/state/output (I/S/O) representation

$$\mathscr{B} = \left\{ w = \Pi \begin{bmatrix} u \\ y \end{bmatrix} \mid \exists x \in (\mathbb{R}^n)^{\mathbb{N}}, \begin{bmatrix} \sigma x \\ y \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} x \\ u \end{bmatrix} \right\}$$

Linearity and time-invariance are naturally defined in terms of  $\mathcal{B}$ 

 $\mathscr{B}$  is linear system  $\iff \mathscr{B}$  is subspace

 $\mathscr{B}$  is time-invariant  $\iff \sigma^{\tau}\mathscr{B} := \mathscr{B}$  for all  $\tau$ 

$$\sigma \mathscr{B} = \big\{ \sigma \mathsf{W} \mid \mathsf{W} \in \mathscr{B} \big\}$$

 $\mathcal{L}^q$  — set of LTI systems with q variables

Equivalence of representations and transformations among them



exercise 3 — from I/S/O to KER representation

# How to check if $w \in \mathscr{B}$ ?

depends on what representation of  $\mathcal B$  is used

different repr. leads to different methods

for example

▶ if *B* is specified by vector difference equation

$$w \in \mathscr{B} \quad \iff \quad R_0 w + R_1 \sigma w + \dots + R_\ell \sigma^\ell w = 0$$

▶ if *B* is specified by input/state/output representation

$$\boldsymbol{w} \in \mathscr{B} \quad \iff \quad \exists \ \boldsymbol{x} \in (\mathbb{R}^n)^{\mathbb{N}}, \ \begin{bmatrix} \sigma \boldsymbol{x} \\ \boldsymbol{y} \end{bmatrix} = \begin{bmatrix} \boldsymbol{A} & \boldsymbol{B} \\ \boldsymbol{C} & \boldsymbol{D} \end{bmatrix} \begin{bmatrix} \boldsymbol{x} \\ \boldsymbol{u} \end{bmatrix}$$

 $w \in \mathscr{B} \iff$  system of linear equations

you have to derive them once

1. using I/S/O representation exercise 1

2. using kernel representation exercise 4

The finite-horizon behavior  $\mathscr{B}|_L$  is used for both analysis and computations

restriction of w to finite interval [1, L]

$$|w|_L := (w(1), \ldots, w(L)) \in (\mathbb{R}^q)^L$$

restriction of  $\mathscr{B}$  to [1, L]

$$\mathscr{B}|_L := \{ w|_L \mid w \in \mathscr{B} \} \subset (\mathbb{R}^q)^L$$

if  $\mathscr{B}$  is linear,  $\mathscr{B}|_L$  is a subspace of  $(\mathbb{R}^q)^L$ 

 $\mathscr{B}|_L$  can be obtained experimentally by collecting "informative" data

collect  $N \ge qL$  random trajectories

$$w_d^1, \ldots, w_d^N \in \mathscr{B}|_L$$

by the linearity of  $\mathcal{B}$ , we have

span { 
$$w_d^1, \ldots, w_d^N$$
 }  $\subseteq \mathscr{B}|_L$ 

with probability one equality holds
Discrete-time LTI systems over finite horizon can be studied using linear algebra only

$$\underbrace{\begin{bmatrix} w_d^1 & \cdots & w_d^N \end{bmatrix}}_{W} \in \mathbb{R}^{qL \times N} - \text{``trajectory matrix''}$$

 $\widehat{\mathscr{B}}|_L = \text{image } W$  — "data-driven model" of  $\mathscr{B}|_L$ 

now, we can do explorations using Matlab

## What is the dimension of $\mathscr{B}|_L$ ? take a random LTI system

m = 2; p = 5; n = 20; B = drss(n, p, m);

#### generate qL random trajectories of length L

```
L = 100; q = m + p; W = []; vec = @(a) a(:);
for i = 1:q*L
    u = rand(L, m); xini = rand(n, 1);
    y = lsim(B, u, [], xini);
    w = [u y]; W = [W vec(w')];
end
```

assuming that image  $W = \mathscr{B}|_L$ , find dim  $\mathscr{B}|_L$ 

for t = 1:L, d(t) = rank(W(1:q\*t, :)); end
stem(d)

## dim $\mathscr{B}|_L$ is a piecewise affine function of L



in particular, dim  $\mathscr{B}|_L = mL + n$ , for all  $L \ge \ell$ 

The set of linear time-invariant systems  $\mathscr{L}$  has structure characterized by set of integers

the dimension of  $\mathscr{B}\in\mathscr{L}$  is determined by

 $\mathbf{m}(\mathscr{B})$  — number of inputs

 $\ell(\mathscr{B})$  — lag (= observability index)

 $\mathbf{n}(\mathscr{B})$  — order (= minimal state dimension)

exercise 2 — find  $\ell(\mathscr{B})$  for given  $\mathscr{B}$ exercise 6 — find  $\mathbf{m}(\mathscr{B}), \ell(\mathscr{B}), \mathbf{n}(\mathscr{B})$  from  $w_{d} \in \mathscr{B}|_{T_{d}}$   $\mathscr{B}_1$  less complex than  $\mathscr{B}_2 \iff \mathscr{B}_1 \subset \mathscr{B}_2$ 

#### in the LTI case, complexity $\leftrightarrow$ dimension

complexity: (# inputs, order, lag)  $c(\mathscr{B}) := (m(\mathscr{B}), n(\mathscr{B}), \boldsymbol{\ell}(\mathscr{B}))$ 

 $\mathscr{L}_c$  — bounded complexity LTI model class

$$\mathscr{L}^{\boldsymbol{q}}_{\boldsymbol{c}} := \{\mathscr{B} \in \mathscr{L}^{\boldsymbol{q}} \mid \boldsymbol{\mathsf{C}}(\mathscr{B}) \leq \boldsymbol{c}\}$$

## Finite vs infinite dimensional LTI systems

 $\mathscr{B} \in \mathscr{L}^q$  finite-dimensional :  $\iff$ 

$$\mathbf{m}(\mathscr{B}) < q$$
  
 $\mathbf{n}(\mathscr{B}) < \infty$ 

### equivalently

- $\mathscr{B}$  has bounded complexity  $\mathbf{c}(\mathscr{B})$
- ► *B* admits rational transfer function representation

parametric representations of  $\mathscr{B} \in \mathscr{L}^q_c$ 

## Summary

$$w \in (\mathbb{R}^q)^{\mathscr{T}}$$
 — *q*-variate signal

 $\mathscr{B} \in \mathscr{L}^q - q$ -variate LTI system

 $\dim \mathscr{B}|_{L} = \mathbf{m}(\mathscr{B})L + \mathbf{n}(\mathscr{B}), \text{ for all } L \geq \boldsymbol{\ell}(\mathscr{B})$ 

exercise 1 — state-space proof of the formula

## Initial conditions specified by "past" trajectory

 $W = W_{\rm p} \wedge W_{\rm f}$ 



#### exercise 23 — dealing with nonzero initial conditions

How long should  $w_p$  be in order to specify the initial conditions for  $w_f$ ?

answer: at least  $\ell(\mathscr{B})$  samples

in general, there are infinitely many  $w_p$ 's that specify the same initial condition

 $w_p$  is a non-minimal state vector

Input/output partitioning of the variables

### $w =: \Pi \begin{bmatrix} u \\ y \end{bmatrix}$ , with $\Pi$ permutation, such that

*u* is input := free variable *y* is output := uniquely defined by  $\mathscr{B}$ ,  $w_{ini}$ , and *u* 

simulation problem:  $(\mathscr{B}, w_{ini}, u) \mapsto y$ 

section 4 of the exercises

parametrization of w by u and wini

## Finding initial conditions (observer)

given  $\mathscr{B}$  and  $w_{f} \in \mathscr{B}|_{T_{f}}$ , find  $w_{p} \in (\mathbb{R}^{q})^{T_{p}}$ , s.t.

$$w_{p} \wedge w_{f} \in \mathscr{B}|_{\mathcal{T}_{p}+\mathcal{T}_{f}}$$

exercise 23 — finding initial conditions

feasibility problem, solution always exists (why?)

in general, it is not unique (is this an issue?)

Initial conditions estimation (smoothing)

given  $\mathscr{B}$  and  $w_{f} \in (\mathbb{R}^{q})^{T_{f}}$ , find  $w_{p} \in (\mathbb{R}^{q})^{T_{p}}$  that

minimize over  $\widehat{w}_{p}, \widehat{w}_{f} || w_{f} - \widehat{w}_{f} ||$ subject to  $\widehat{w}_{p} \wedge \widehat{w}_{f} \in \mathscr{B}|_{\mathcal{T}_{p} + \mathcal{T}_{f}}$ section 6 of the exercises

as byproduct we find "smoothed" trajectory  $\widehat{w}_{f}$ 

errors-in-variables (EIV) smoother

## Projection on *B*

given  $\mathscr{B}$  and  $w \in (\mathbb{R}^q)^T$ , find  $\widehat{w} \in (\mathbb{R}^q)^T$  that

minimize over  $\widehat{w} \| w - \widehat{w} \|$ subject to  $\widehat{w} \in \mathscr{B}|_{\mathcal{T}}$ 

#### equivalent to the EIV smoothing problem

#### prior knowledge about the initial conditions

- completely unknown
- uncertain (mean value and covariance are given)
- given exactly

# Most powerful unfalsified model of $\mathscr{B}_{mpum}(w_d)$

### exact identification problem



#### multi-objective optimization problem

- complexities are compared in the lexicographic order
- more inputs imply higher complexity irrespective of order

feasibility and uniqueness are guaranteed

$$\mathscr{B}_{mpum}(w_d) := span\{w_d, \sigma w_d, \sigma^2 w_d, \dots\}$$

There is a problem with  $\mathscr{B}_{mpum}(w_d)$ in case of finite data  $w_d \in (\mathbb{R}^q)^{T_d}$ 

 $\widehat{\mathscr{B}} := \mathscr{B}_{mpum}(w_d)$  is autonomous exercise 5

solution: impose the upper bound

$$\boldsymbol{\ell}(\widehat{\boldsymbol{\mathscr{B}}}) \leq \ell_{\max} := \left\lfloor \frac{T_{\mathsf{d}} + 1}{q+1} \right\rfloor - 1$$

exact identification —  $\mathscr{B}_{mpum}(w_d)$  computation exercise 7 — find kernel repr. of  $\mathscr{B}_{mpum}(w_d)$ 

## Summary

"past" trajectory — specifies initial conditions simulation: with  $w =: \prod \begin{bmatrix} u \\ y \end{bmatrix}$ ,  $(\mathscr{B}, w_{ini}, u) \mapsto y$ inverse problem:  $w_d \mapsto \mathscr{B}_{mpum}(w_d)$  More system properties

controllability

autonomy

stability

What means that  $\mathcal{B}$  is controllable?

controllability is the property of "patching" any past trajectory with any future trajectory

 $W_{\rm p} \wedge W_{\rm c} \wedge W_{\rm f} \in \mathscr{B}$ 



Compare with the classical definition: transfer from any initial to any terminal state

property of a state-space representation of  ${\mathscr B}$ 

- is lack of controllability due to a "bad" choice of the state or due to an intrinsic issue with the system?
- in the LTI case, does it make sense to talk about controllability of a transfer function representation?
- how to quantify the "distance" to uncontrollability?

does not apply to infinite dimensional system

## Methods for checking controllability

how to check controllability of an LTI system?

using state-space representation:

- 1. ensure minimality (in the behavioral sense)
- 2. perform rank test for the controllability matrix

using matrix fraction representation:

$$\mathscr{B} = \left\{ w = \Pi \left[ \begin{smallmatrix} u \\ y \end{smallmatrix} 
ight] \in (\mathbb{R}^q)^{\mathbb{N}} \mid N(\sigma)u = D(\sigma)y 
ight\}$$

### $\mathscr{B}$ autonomous $\iff \mathscr{B}$ has no inputs

autonomy: most extreme uncontrollability

any system has decomposition

$$\mathscr{B} = \mathscr{B}_{controllable} + \mathscr{B}_{autonomous}$$

 $\mathscr{B} \in \mathscr{L}^q$  and autonomous if and only if

 $w \in \mathscr{B}$  is sum of polynomials times exponentials

Stability is naturally property of the behavior

 $\mathscr{B}$  stable  $\iff w(t) \rightarrow 0$  as  $t \rightarrow \infty$ , for all  $w \in \mathscr{B}$ 

stability implies autonomy

 $\mathscr{B} \in \mathscr{L}^q$  and stable if and only if  $w \in \mathscr{B}$  converges exponentially to 0

## Summary

controllability: patching past/future trajectories

autonomy: no inputs ( $\mathbf{m}(\mathscr{B}) = 0$ )

decomposition into controllable and autonomous

 *B* ∈ L<sup>q</sup> autonomous ⇔ w = Σ<sup>n</sup><sub>i=1</sub> polynomial<sub>i</sub> × exp<sub>λi</sub>

 λ<sub>1</sub>,...,λ<sub>n</sub> — poles of the system B

stability:  $w(t) \rightarrow 0$  as  $t \rightarrow \infty$ , for all  $w \in \mathscr{B}$ 

BIBO stability is not a property of *B* 



Introduction: the need

#### Basics: notation and conventions

Data-driven interpolation and approximation

The new "data-driven" paradigm obtains desired solution directly from given data



Data-driven does not mean model-free

data-driven problems do assume model

however, specific representation is not fixed

the methods we review are non-parametric

Data-driven representation (infinite horizon)

#### data: exact infinite trajectory $w_d$ of $\mathscr{B} \in \mathscr{L}$

$$\widehat{\mathscr{B}} = \mathscr{B}_{mpum}(w_d) = span\{w_d, \sigma w_d, \sigma^2 w_d, \dots\}$$

identifiability condition:  $\mathscr{B} = \widehat{\mathscr{B}}$ 

Consecutive application of  $\sigma$  on finite  $w_d$  results in Hankel matrix with missing values

for 
$$w_d = (w_d(1), \dots, w_d(T_d))$$
 and  $1 \le L \le T_d$   
 $\mathscr{H}_L(w_d) := \left[ (\sigma^0 w_d) |_L (\sigma^1 w_d) |_L \cdots (\sigma^{T_d - L} w_d) |_L \right]$ 

Data-driven representation (finite horizon)

the finite horizon data-driven representation

$$\mathscr{B}|_L = \widehat{\mathscr{B}}|_L := \text{image } \mathscr{H}_L(w_d)$$
 (DD-REPR)

holds if and only if

$$\operatorname{rank} \mathscr{H}_L(w_d) = L\mathbf{m}(\mathscr{B}) + \mathbf{n}(\mathscr{B})$$
 (GPE)

GPE — generalized persistency of excitation

exercise 1 — from I/S/O representation to  $\mathscr{B}|_L$ 

Identifiability condition verifiable from  $w_d \in \mathscr{B}|_{T_d}$  and  $(m, \ell, n)$ 

### fact: $\mathscr{B} = \mathscr{B}' \iff \mathscr{B}|_{\ell+1} = \mathscr{B}'|_{\ell+1}$ , then

$$\widehat{\mathscr{B}} = \mathscr{B} \quad \Longleftrightarrow \quad \widehat{\mathscr{B}}|_{\ell+1} = \mathscr{B}|_{\ell+1} \\ \Leftrightarrow \quad \dim \widehat{\mathscr{B}}|_{\ell+1} = \dim \mathscr{B}|_{\ell+1}$$

 $\mathscr{B}$  is identifiable from  $w_d \in \mathscr{B}|_{\mathcal{T}_d}$  if and only if

$$\operatorname{rank} \mathscr{H}_{\ell+1}(w_d) = (\ell+1)m + n$$

The "fundamental lemma" is an input design result

J.C. Willems et al., A note on persistency of excitation Systems & Control Letters, (54)325–329, 2005

### sufficient conditions for (DD-REPR)

1. 
$$W_{d} = \begin{bmatrix} u_{d} \\ y_{d} \end{bmatrix}$$

2. B controllable

3.  $\mathscr{H}_{L+n}(u_d)$  full row rank

(PE)

### PE — persistency of excitation

Generic data-driven problem: trajectory interpolation/approximation

"data trajectory"  $W_{d} \in \mathscr{B}|_{\mathcal{T}_{d}}$ and elements given: of a trajectory  $W \in \mathscr{B}|_{\mathcal{T}}$ 

 $(w|_{I_{\text{given}}}$  selects the elements of w, specified by  $I_{\text{given}}$ )

W

minimize over  $\widehat{w} \| w \|_{I_{\text{diven}}} - \widehat{w} \|_{I_{\text{diven}}} \|$ aim: subject to  $\widehat{w} \in \mathscr{B}|_{\mathcal{T}}$ 

$$\widehat{\boldsymbol{w}} = \mathscr{H}_{T}(\boldsymbol{w}_{d}) \big( \mathscr{H}_{T}(\boldsymbol{w}_{d}) |_{I_{\text{given}}} \big)^{+} \boldsymbol{w} |_{I_{\text{given}}} \quad \text{(SOL}$$

## Special cases

#### simulation

section 4

- given data: initial condition and input
- to-be-found: output (exact interpolation)

### smoothing

- given data: noisy trajectory
- to-be-found: l<sub>2</sub>-optimal approximation

### tracking control

- given data: to-be-tracked trajectory
- to-be-found:  $\ell_2$ -optimal approximation

### sections 6 and 7

section 8

## Generalizations

multiple data trajectories  $w_d^1, \ldots, w_d^N$ 

$$\widehat{\mathscr{B}}|_{L} = \text{image} \underbrace{ \left[ \mathscr{H}_{L}(w_{d}^{1}) \cdots \mathscr{H}_{L}(w_{d}^{N}) \right] }_{\mathcal{H}_{L}(w_{d}^{N})}$$

mosaic-Hankel matrix

#### w<sub>d</sub> not exact / noisy

mini-projects

maximum-likelihood estimation → Hankel structured low-rank approximation/completion nuclear norm and ℓ<sub>1</sub>-norm relaxations → nonparametric, convex optimization problems

#### nonlinear systems

mini-projects

results for special classes of nonlinear systems: Volterra, Wiener-Hammerstein, bilinear, ... Summary: data-driven signal processing

#### data-driven representation

leads to general, simple, practical methods

interpolation/approximation of trajectories

simulation, filtering and control are special cases assumes only LTI dynamics; no hyper parameters

dealing with noise and nonlinearities

nonlinear optimization convex relaxations

## The data w<sub>d</sub> being exact vs inexact / "noisy"

### $w_d$ exact and satisfying (GPE)

- "systems theory" problems
- image  $\mathcal{H}_L(w_d)$  is nonparametric finite-horizon model
- data-driven solution = model-based solution

#### $w_d$ inexact, due to noise and/or nonlinearities

- naive approach: apply the solution (SOL) for exact data
- ▶ rigorous: assume noise model ~→ ML estimation problem
- heuristics: convex relaxations of the ML estimator
The maximum-likelihood estimation problem in the errors-in-variables setup is nonconvex

errors-in-variables setup:  $w_d = \overline{w}_d + \widetilde{w}_d$ 

ML problem: given  $w_d$ , c, and  $w|_{I_{given}}$ 

$$\begin{array}{ll} \underset{g}{\text{minimize}} & \|w\|_{l_{\text{given}}} - \mathscr{H}_{T}(\widehat{w}_{d}^{*})\|_{l_{\text{given}}}g\| \\ \text{subject to} & \widehat{w}_{d}^{*} = \arg\min_{\widehat{w}_{d},\widehat{\mathscr{B}}} & \|w_{d} - \widehat{w}_{d}\| \\ & \text{subject to} & \widehat{w}_{d} \in \widehat{\mathscr{B}}|_{\mathcal{T}_{d}} \text{ and } \widehat{\mathscr{B}} \in \mathscr{L}_{c}^{q} \end{array}$$

# The ML estimation problem is equivalent to Hankel structured low-rank approximation

$$\begin{array}{ll} \underset{g}{\text{minimize}} & \|w\|_{I_{\text{given}}} - \mathscr{H}_{T}(\widehat{w}_{d}^{*})\|_{I_{\text{given}}}g\|\\ \text{subject to} & \widehat{w}_{d}^{*} = \arg\min_{\widehat{w}_{d},\widehat{\mathscr{B}}} & \|w_{d} - \widehat{w}_{d}\|\\ & \text{subject to} & \widehat{w}_{d} \in \widehat{\mathscr{B}}|_{T_{d}} \text{ and } \widehat{\mathscr{B}} \in \mathscr{L}_{C}^{q}\\ & \uparrow \end{array}$$

 $\begin{array}{ll} \underset{g}{\text{minimize}} & \|w\|_{l_{\text{given}}} - \mathscr{H}_{T}(\widehat{w}_{d}^{*})\|_{l_{\text{given}}}g\| \\ \text{subject to} & \widehat{w}_{d}^{*} = \arg\min_{\widehat{w}_{d}} & \|w_{d} - \widehat{w}_{d}\| \\ & \text{subject to} & \operatorname{rank}\mathscr{H}_{\ell+1}(\widehat{w}_{d}) \leq (\ell+1)m + n \end{array}$ 

# Solution methods

## local optimization

- choose a parametric representation of  $\widehat{\mathscr{B}}(\theta)$
- optimize over  $\widehat{w}$ ,  $\widehat{w_{d}}$ , and  $\theta$
- depends on the initial guess

#### convex relaxation based on the nuclear norm

$$\begin{array}{ll} \text{minimize} \quad \text{over } \widehat{w}_{\mathsf{d}} \text{ and } \widehat{w} & \|w|_{l_{\mathsf{given}}} - \widehat{w}|_{l_{\mathsf{given}}}\| + \|w_{\mathsf{d}} - \widehat{w}_{\mathsf{d}}\| \\ & + \gamma \cdot \left\| \begin{bmatrix} \mathscr{H}_{\Delta}(\widehat{w}_{\mathsf{d}}) & \mathscr{H}_{\Delta}(\widehat{w}) \end{bmatrix} \right\|_{*} \end{array}$$

convex relaxation based on  $\ell_1$ -norm (LASSO) minimize over  $g ||w|_{I_{given}} - \mathscr{H}_T(w_d)|_{I_{given}}g|| + \lambda ||g||_1$ 

# Empirical validation on real-life datasets

	data set name	$T_{d}$	т	р
1	Air passengers data	144	0	1
2	Distillation column	90	5	3
3	pH process	2001	2	1
4	Hair dryer	1000	1	1
5	Heat flow density	1680	2	1
6	Heating system	801	1	1

G. Box, and G. Jenkins. Time Series Analysis: Forecasting and Control, Holden-Day, 1976

B. De Moor, et al.DAISY: A database for identification of systems. Journal A, 38:4–5, 1997

 $\ell_1$ -norm regularization with optimized  $\lambda$  achieves the best performance

$$e_{\mathsf{missing}} \coloneqq rac{\|w|_{I_{\mathsf{missing}}} - \widehat{w}|_{I_{\mathsf{missing}}}\|}{\|w|_{I_{\mathsf{missing}}}\|} \ 100\%$$

	data set name	naive	ML	LASSO
1	Air passengers data	3.9	fail	3.3
2	Distillation column	19.24	17.44	9.30
3	pH process	38.38	85.71	12.19
4	Hair dryer	12.35	8.96	7.06
5	Heat flow density	7.16	44.10	3.98
6	Heating system	0.92	1.35	0.36

# Tuning of $\lambda$ and sparsity of *g* (datasets 1, 2)



# Tuning of $\lambda$ and sparsity of *g* (datasets 3, 4)



# Tuning of $\lambda$ and sparsity of *g* (datasets 5, 6)



## Summary: convex relaxations

#### w<sub>d</sub> exact ~> systems theory

- exact analytical solution
- current work: efficient real-time algorithms

#### w<sub>d</sub> inexact ~> nonconvex optimization

- subspace methods
- Iocal optimization
- convex relaxations

#### empirical validation

- the naive approach works (surprisingly) well
- parametric local optimization is not robust
- ℓ<sub>1</sub>-norm regularization gives the best results



Constructive proof of the fundamental lemma

Pedagogical example: Free fall prediction

Case study: Dynamic measurement

Nonparametric frequency response estimation

Generalization for nonlinear systems



## Constructive proof of the fundamental lemma

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The fundamental lemma gives data-driven finite horizon representation of LTI system  $\mathscr{B}$ 

## $\mathscr{B}|_L = \operatorname{image} \mathscr{H}_L(w_d)$ (DD-REPR)

#### assumptions:

- A0  $w_d = \begin{bmatrix} u_d \\ y_d \end{bmatrix}$  is a trajectory of an LTI system  $\mathscr{B}$ A1  $\mathscr{B}$  is controllable
- A2  $u_d$  is persistently exciting of order L + n

Decoding the notation  $\mathscr{B}|_L = \operatorname{image} \mathscr{H}_L(w_d)$ 

*B* — system's behavior, *i.e.*, set of trajectories

 $\mathscr{B}|_L$  — restriction of  $\mathscr{B}$  to the interval [1, L]

 $w_d := (w_d(1), \dots, w_d(T_d))$  — "data" trajectory

$$\mathscr{H}_{L}(w_{d}) := \begin{bmatrix} w_{d}(1) & w_{d}(2) & \cdots & w_{d}(T_{d}-L+1) \\ \vdots & \vdots & & \vdots \\ w_{d}(L) & w_{d}(L+1) & \cdots & w_{d}(T_{d}) \end{bmatrix}$$

 $PE(u_d) := \max L$ , such that  $\mathscr{H}_L(u_d)$  is f.r.r.

We address the following issues/questions

## proof by contradiction

What is the meaning/interpretation of the conditions?

## sufficiency of the conditions

How conservative are they? Can they be improved?

#### conjecture

The extra PE of order n is generically not needed. What are the nongeneric cases when it is needed?

## Answers

#### constructive proof in the single-input case

$$\mathsf{PE}(u_{\mathsf{d}}) = n_u \iff u_{\mathsf{d}} \in \mathscr{B}_u|_{\mathcal{T}_{\mathsf{d}}}, \text{ where } \mathscr{B}_u \text{ is autonomous LTI of order } n_u$$

#### shows that the FL is nonconservative conjecture: it is conservative in the multi-input case

characterizes the nongeneric cases they correspond to special initial conditions Necessary and sufficient condition for the data-driven representation

$$\operatorname{rank} \mathscr{H}_L(w_d) = mL + n,$$
 (GPE)

nonconservative (necessary and sufficient) general no I/O partitioning and controllability verifiable from  $w_d$  with prior knowledge of (m, n) The fundamental lemma is input design result

#### input design problem

choose  $u_d$ , so that (DD-REPR) holds for any initial cond.

## refined problem statement

find nonconservative conditions on  $u_d$  and  $\mathcal{B}$ , under which

for  $\forall w_{d,ini}, w_{d,ini} \land w_{d} \in \mathscr{B}|_{\mathcal{T}_{ini} + \mathcal{T}_{d}}$  satisfies (GPE) (GOAL)

subproblem: find  $w_{ini}$  that minimize rank  $\mathscr{H}_L(w_d)$ 

# Obvious necessary conditions

A0: exact representation requires exact data and input design requires input/output partition

- A1: for uncontrollable  $\mathscr{B} = \mathscr{B}_{ctr} \oplus \mathscr{B}_{aut}$ 

  - w<sub>d,aut</sub> is completely determined by w<sub>d,ini</sub>
  - there is  $w_{d,ini}$ , such that  $w_{d,aut} = 0 \implies$  (GPE) doesn't hold

#### A2': $u_d$ is persistently exciting of order L

- since *u* is an input,  $\Pi_u \mathscr{B}|_L = \mathbb{R}^{\mathbf{m}(\mathscr{B})L}$
- ▶ for (GPE) to hold true, image  $\mathscr{H}_L(u_d) = \mathbb{R}^{\mathbf{m}(\mathscr{B})L}$
- equivalently,  $\mathscr{H}_L(u_d)$  must be full row-rank

Find the minimal *k*, such that (GOAL) holds under A0, A1, and  $PE(u_d) = L + k$ 

first, we solve the subproblem find  $w_{ini}^*$  that minimize rank  $\mathscr{H}_L(w_d)$ 

then, we check (GPE) for  $w_{ini}^*$ 

 $\rightsquigarrow$  minimal  $k \implies$  nonconservative PE condition

The PE condition is equivalent to existence of an LTI input model

$$u_{\mathsf{d}} \in (\mathbb{R})^{T_{\mathsf{d}}}$$
 and  $\mathsf{PE}(u_{\mathsf{d}}) = n_{u}$ 

 $u_{d} \in \mathscr{B}_{u}|_{T_{d}}$  — autonomous LTI,  $T_{d} \ge 2n_{u} - 1$  $\mathscr{B}_{u} = \mathscr{B}_{ss}(A_{u}, C_{u})$  with  $(A_{u}, x_{u, ini})$  controllable



# Augmented system with the input model

$$\mathscr{B}_{\text{ext}} = \mathscr{B}_{\text{ss}}(A_{\text{ext}}, C_{\text{ext}}), \text{ with } x_{\text{ext}} = \begin{bmatrix} x_{u} \\ x \end{bmatrix}$$

$$A_{\text{ext}} = \begin{bmatrix} A_u & 0 \\ BC_u & A \end{bmatrix} \quad C_{\text{ext}} = \begin{bmatrix} C_u & 0 \\ DC_u & C \end{bmatrix}$$

$$\mathcal{B}_{\text{ext}} = \mathcal{B}_{\text{ss}} \left( \mathbf{A}_{\text{ext}}', \mathbf{C}_{\text{ext}}' \right), \text{ where } \mathbf{x}_{\text{ext}}' = \begin{bmatrix} \mathbf{x}_{u} \\ \mathbf{V}\mathbf{x}_{u} + \mathbf{x} \end{bmatrix}$$
$$\mathbf{A}_{\text{ext}}' = \begin{bmatrix} \mathbf{A}_{u} & \mathbf{0} \\ \mathbf{0} & \mathbf{A} \end{bmatrix}, \quad \mathbf{C}_{\text{ext}}' = \begin{bmatrix} \mathbf{C}_{u} & \mathbf{0} \\ \mathbf{C}' & \mathbf{C} \end{bmatrix}, \quad \mathbf{C}' := \mathbf{D}\mathbf{C}_{u} - \mathbf{C}\mathbf{V}$$

V is solution of the Sylvester equation  $AV - VA_u = BC_u$ 

The nongeneric cases correspond to special initial conditions  $x_{ini} = -Vx_{u,ini}$ 

#### which eliminates from $w_d$ the transient due to $\mathscr{B}$

then, rank  $\mathscr{H}_L(w_d) \leq \mathsf{PE}(u_d) = n_u$ 

next, we show that rank  $\mathscr{H}_L(w_d) = n_u$ 

assume simple eigenvalues  $\lambda_{u,1}, \ldots, \lambda_{u,n_u}$  of  $\mathscr{B}_u$ 

$$u_{\mathsf{d}} = \sum_{i=1}^{n_u} a_i \exp_{\lambda_{u,i}}$$

assume simple eigenvalues  $\lambda_1, \ldots, \lambda_n$  of  $\mathscr{B}$ 

$$y_{d} = \sum_{i=1}^{n_{u}} b_{i} \exp_{\lambda_{u,i}} + \underbrace{\sum_{j=1}^{n} c_{j} \exp_{\lambda_{j}}}_{\text{transient}}$$

• 
$$b_i = H(e^{i\lambda_{u,i}})a_i$$
, where  $H(z) := C(Iz - A)^{-1}B + D$   
•  $w_{ini} = w_{ini}^* \implies c_j = 0$ 

## using Vandermonde matrix, we rewrite $(u_d, y_d)$

$$u_{d} = \underbrace{\begin{bmatrix} \lambda_{u,1}^{1} & \cdots & \lambda_{u,n_{u}}^{1} \\ \vdots & & \vdots \\ \lambda_{u,1}^{T_{d}} & \cdots & \lambda_{u,n_{u}}^{T_{d}} \end{bmatrix}}_{V_{T_{d}}(\lambda_{u})} \underbrace{\begin{bmatrix} a_{1} \\ \vdots \\ a_{n_{u}} \end{bmatrix}}_{a} = V_{T_{d}}(\lambda_{u})a$$

and

$$y_{d} = V_{T_{d}}(\lambda_{u}) \underbrace{ \begin{bmatrix} H(e^{i\lambda_{u,1}}) & & \\ & \ddots & \\ & H(e^{i\lambda_{u,n_{u}}}) \end{bmatrix}}_{H(\lambda_{u})} \begin{bmatrix} a_{1} \\ \vdots \\ a_{n_{u}} \end{bmatrix}$$
$$= V_{T_{d}}(\lambda_{u}) \underbrace{H(\lambda_{u})a}_{b} = V_{T_{d}}(\lambda_{u})b$$

then, for  $w_d$ , we obtain

$$w_{\mathsf{d}} = \Pi_{\mathcal{T}_{\mathsf{d}}} \begin{bmatrix} V_{\mathcal{T}_{\mathsf{d}}}(\lambda_{u}) \\ V_{\mathcal{T}_{\mathsf{d}}}(\lambda_{u})H(\lambda_{u}) \end{bmatrix} a$$

 $\Pi_{\mathcal{T}_d} \in \mathbb{R}^{2\mathcal{T}_d \times 2\mathcal{T}_d} \text{ permutation, such that } w_d = \Pi_{\mathcal{T}_d} \begin{bmatrix} u_d \\ y_d \end{bmatrix}$ 

finally, the Hankel matrix is expressed as  

$$\mathscr{H}_{L}(w_{d}) = \underbrace{\prod_{L} \begin{bmatrix} V_{L}(\lambda_{u}) \\ V_{L}(\lambda_{u})H(\lambda_{u}) \end{bmatrix}}_{W_{L}} \underbrace{\begin{bmatrix} a \quad \Lambda_{u}a \quad \Lambda_{u}^{2}a \quad \cdots \quad \Lambda_{u}^{T_{d}-L}a \end{bmatrix}}_{\text{controllability matrix of } (\Lambda_{u}, a)}$$

$$\Lambda_{u} := \operatorname{diag}(\lambda_{u,1}, \dots, \lambda_{u,n_{u}})$$

 $(\Lambda_u, a)$  is controllable because  $PE(u_d) = n_u$ 

1.  $a_i \neq 0$  for all *i* 2.  $\lambda_{u,i} \neq \lambda_{u,j}$  for all  $i \neq j$ 

for  $k \leq n$ ,  $W_L$  is full column rank

with W<sub>L</sub> = [w<sup>1</sup> ... w<sup>n<sub>u</sub></sup>], w<sup>i</sup> are trajectories (w<sup>i</sup> ∈ ℬ|<sub>L</sub>)
 λ<sub>u,i</sub> ≠ λ<sub>u,j</sub> for all i ≠ j ⇒ independent responses

rank 
$$\mathscr{H}_{L}(w_{d}) = \begin{cases} L+k, & \text{for } k = 1, \dots, n \\ L+n, & \text{for } k = n+1, \dots \end{cases}$$

k = n is the minimal value for (GPE) to hold

## Comments

the zeros of  $\mathscr{B}$  don't play role in the analysis

simple eigenvalues assumptions can be relaxed

"robustifying" the conditions

exact condition:robust version: $a_i \neq 0$ , for all i $a_i > \varepsilon$  $\lambda_{u,i} \neq \lambda_{u,j}$ , for all  $i \neq j$ the  $\lambda_{u,i}$ 's are "well spread"

conjecture: in multi-input case, A2 can be tightened,  $PE(u_d) = n + \text{controllability index } \mathscr{B}$ 



Constructive proof of the fundamental lemma

Pedagogical example: Free fall prediction

Case study: Dynamic measurement

Nonparametric frequency response estimation

Generalization for nonlinear systems

The goal is to predict free fall trajectory without knowing the laws of physics

object with mass m, falling in gravitational field

- ► *y* position
- $\blacktriangleright$   $v := \dot{y}$  velocity
- y(0), v(0) initial condition

task: given initial condition, find the trajectory y

model-based approach:

1. physics  $\mapsto$  model 2. model + ini. cond.  $\mapsto$  y

► data-driven approach: data  $y_d^1, \ldots, y_d^N$  + ini. cond.  $\mapsto y$ 

# Modeling from first principles leads to affine time-invariant state-space model

second law of Newton + the law of gravity

$$m\ddot{y} = m\begin{bmatrix} 0\\ 9.81\end{bmatrix} + f$$
, where  $y(0) = y_{ini}$  and  $\dot{y}(0) = v_{ini}$ 

• 9.81 — gravitational constant •  $f = -\alpha y -$  force due to friction in the

•  $f = -\gamma v$  — force due to friction in the air

state  $x := (y_1, \dot{y}_1, y_2, \dot{y}_2, x_5)$ , where  $x_5 = -9.81$ 

initial state  $x_{ini} := (y_{ini,1}, v_{ini,1}, y_{ini,2}, v_{ini,2}, -9.81)$ 

Modeling from first principles leads to affine time-invariant state-space model

$$\dot{x} = \begin{bmatrix} 0 & 1 & & & \\ 0 & -\gamma/m & & & \\ & 0 & 1 & & \\ & 0 & -\gamma/m & 1 \\ & & & 0 \end{bmatrix} x, \qquad x(0) = \begin{bmatrix} y_{\text{ini},1} \\ v_{\text{ini},2} \\ v_{\text{ini},2} \\ -9.81 \end{bmatrix}$$
$$y = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \end{bmatrix} x$$

data: N, T-samples long discretized trajectories

# Simulation setup and data

write a function fall that simulates free fall
y = fall(y0, v0, t, m, gamma)

simulate N=10, T=100-samples long trajectories

m = 1; gamma = 0.5; N = 10; T = 100; t = linspace(0, 1, T); for i = 1:N, y{i} = fall(rand(2,1), rand(2,1), t,gamma,m); end

## and to-be-predicted trajectory

y\_new = fall(rand(2,1), rand(2,1), t,gamma,m);

# Data-driven free fall prediction method

data "informativity" condition:

$$\operatorname{rank}\underbrace{\begin{bmatrix} y_{d}^{1} & \cdots & y_{d}^{N} \end{bmatrix}}_{D} = 5$$

algorithm for data-driven prediction:

1. solve 
$$\begin{bmatrix} y_{d}^{1}(1) & \cdots & y_{d}^{N}(1) \\ y_{d}^{1}(2) & \cdots & y_{d}^{N}(2) \\ y_{d}^{1}(3) & \cdots & y_{d}^{N}(3) \end{bmatrix} g = \underbrace{\begin{bmatrix} y(1) \\ y(2) \\ y(3) \end{bmatrix}}_{\text{ini. cond.}}$$

2. define y := Dg

Verify that the data-driven prediction "works"

## check the data "informativity" condition

[rank(D) rank([vec(y\_new') D])] % -> [ 5 5 ]

## implement the data-driven computation method

verify the computed solution

# Summary: prediction of free fall trajectory

## first principles modeling

- use the second law of Newton and the law of gravity
- in particular, the Earth's gravitational constant is used
- lead to an autonomous affine time-invariant system

#### data-driven methods

- bypass the knowledge of the physical laws
- automatically infer and use them
- no hyper-parameters to tune



Constructive proof of the fundamental lemma

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## Case study: Dynamic measurement

Nonparametric frequency response estimation

Generalization for nonlinear systems
My interest in dynamic measurement started from a textbook problem

"A thermometer reading 21°C, which has been inside a house for a long time, is taken outside. After one minute the thermometer reads 15°C; after two minutes it reads 11°C. What is the outside temperature?"

According to Newton's law of cooling, an object of higher temperature than its environment cools at a rate that is proportional to the difference in temperature. Main idea: predict the steady-state value from the first few samples of the transient

#### textbook problem:

- 1st order dynamics
- 3 noise-free samples
- batch solution

#### generalizations:

- $n \ge 1$  order dynamics
- $T \ge 3$  noisy (vector) samples
- recursive computation

#### implementation and practical validation

## Thermometer: first order dynamical system

 $\begin{array}{ccc} \text{environmental} & \xrightarrow{\text{heat transfer}} & \text{thermometer's} \\ \text{temperature } \bar{u} & & \text{reading } y \end{array}$ 

measurement process: Newton's law of cooling

$$y = a(\bar{u} - y)$$

heat transfer coefficient a > 0

## Scale: second order dynamical system



$$(M+m)\frac{\mathrm{d}}{\mathrm{d}\,t}y+dy+ky=g\bar{u}$$

The measurement process dynamics depends on the to-be-measured mass



Dynamic measurement: take into account the dynamical properties of the sensor

to-be-measured variable *u* 

measurement process

measured variable *y* 

assumption 1: measured variable is constant  $u(t) = \bar{u}$ 

assumption 2: the sensor is stable LTI system

assumption 3: sensor's DC-gain = 1 (calibrated sensor)

The data is generated from LTI system with output noise and constant input



assumption 4: e is a zero mean, white, Gaussian noise

using a state space representation of the sensor

$$x(t+1) = Ax(t),$$
  $x(0) = x_0$   
 $y_0(t) = cx(t)$ 

we obtain



## Maximum-likelihood model-based estimator

solve approximately

$$\begin{bmatrix} \mathbf{1}_{T_{\mathsf{d}}} & \mathscr{O}_{T_{\mathsf{d}}} \end{bmatrix} \begin{bmatrix} \widehat{u} \\ \widehat{x}_{\mathsf{0}} \end{bmatrix} \approx y_{\mathsf{d}}$$

standard least-squares problem

minimize over 
$$\hat{y}$$
,  $\hat{u}$ ,  $\hat{x}_0 ||y_d - \hat{y}||$   
subject to  $\begin{bmatrix} \mathbf{1}_{T_d} & \mathcal{O}_{T_d} \end{bmatrix} \begin{bmatrix} \hat{u} \\ \hat{x}_0 \end{bmatrix} = \hat{y}$ 

recursive implementation ~~ Kalman filter

## Subspace model-free method

goal: avoid using the model parameters (A, C,  $\mathcal{O}_{T_d}$ )

in the noise-free case, due to the LTI assumption,

$$\Delta y(t) := y(t) - y(t-1) = y_0(t) - y_0(t-1)$$

satisfies the same dynamics as y<sub>0</sub>, *i.e.*,

$$egin{aligned} x(t+1) &= Ax(t), \qquad x(0) &= \Delta x \ \Delta y(t) &= cx(t) \end{aligned}$$

Hankel matrix—construction of multiple "short" trajectories from one "long" trajectory

$$\mathscr{H}(\Delta y) := \begin{bmatrix} \Delta y(1) & \Delta y(2) & \cdots & \Delta y(n) \\ \Delta y(2) & \Delta y(3) & \cdots & \Delta y(n+1) \\ \Delta y(3) & \Delta y(4) & \cdots & \Delta y(n+2) \\ \vdots & \vdots & \vdots \\ \Delta y(T-n) & \Delta y(T-n) & \cdots & \Delta y(T-1) \end{bmatrix}$$

fact: if rank  $\mathscr{H}(\Delta y) = n$ , then

image 
$$\mathscr{O}_{T-n} = \operatorname{image} \mathscr{H}(\Delta y)$$

#### model-based equation

$$\begin{bmatrix} \mathbf{1}_{T_{\mathsf{d}}} & \mathscr{O}_{T_{\mathsf{d}}} \end{bmatrix} \begin{bmatrix} \bar{u} \\ \widehat{x}_0 \end{bmatrix} = y$$

#### data-driven equation

$$\begin{bmatrix} \mathbf{1}_{T-n} \quad \mathscr{H}(\Delta \mathbf{y}) \end{bmatrix} \begin{bmatrix} \bar{\mathbf{u}} \\ \ell \end{bmatrix} = \mathbf{y}|_{T-n} \qquad (*)$$

#### subspace method

solve (\*) by (recursive) least squares

## **Empirical validation**

dashed		true parameter value $\bar{u}$
solid		true output trajectory y0
dotted		naive estimate $\hat{u} = G^+ y$
dashed	—	model-based Kalman filter
bashed-dotted	—	data-driven method

estimation error:  $e := \frac{1}{N} \sum_{i=1}^{N} \| \bar{u} - \hat{u}^{(i)} \|$ 

(for N = 100 Monte-Carlo repetitions)

## Simulated data of dynamic cooling process



best is the Kalman filter (maximum likelihood estimator)

## Simulation with time-varying parameter



## Proof of concept prototype



## Results in real-life experiment



## Summary

dynamic measurement

steady-state value prediction

#### the subspace method is applicable for

- high order dynamics
- noisy vector observations
- online computation

#### future work / open problems

- numerical efficiency
- real-time uncertainty quantification
- generalization to nonlinear systems

## Outline

Constructive proof of the fundamental lemma

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## Problem formulation

## given: "data" trajectory $(u_d, y_d) \in \mathscr{B}|_{T_d}$ and $z \in \mathbb{C}$

find: H(z), where H is the transfer function of  $\mathscr{B}$ 

Direct data-driven solution we are interested in trajectory

$$w = \begin{bmatrix} u \\ y \end{bmatrix} = \begin{bmatrix} \exp_z \\ \widehat{H} \exp_z \end{bmatrix} \in \mathscr{B}, \text{ where } \exp_z(t) := z^t$$

using the data-driven representation, we have

$$\begin{bmatrix} \mathscr{H}_{\mathsf{L}}(u_{\mathsf{d}}) \\ \mathscr{H}_{\mathsf{L}}(y_{\mathsf{d}}) \end{bmatrix} g = \begin{bmatrix} \mathsf{z} \\ \widehat{H} \mathsf{z} \end{bmatrix}, \quad \text{where } \mathsf{z} := \begin{bmatrix} z^1 \\ \vdots \\ z^{\mathsf{L}} \end{bmatrix}$$

which leads to the system

$$\begin{bmatrix} 0 & \mathscr{H}_{L}(u_{d}) \\ -\mathbf{z} & \mathscr{H}_{L}(y_{d}) \end{bmatrix} \begin{bmatrix} \widehat{H} \\ g \end{bmatrix} = \begin{bmatrix} \mathbf{z} \\ 0 \end{bmatrix}$$
(SYS)

Solution method: solve (SYS) for  $\widehat{H}$ 

under (GPE) with  $L \ge \ell + 1$ ,  $\widehat{H} = H(z)$ 

without prior knowledge of  $\ell$ 

$$\textit{L} = \textit{L}_{max} := \lfloor (\textit{T}_d + 1)/3 \rfloor$$

#### trivial generalization to

- multivariable systems
- multiple data trajectories {  $w_d^1, \ldots, w_d^N$  }
- evaluation of H(z) at multiple points in  $\{z_1, \ldots, z_K\} \in \mathbb{C}^K$

Comparison with classical nonparametric frequency response estimation methods

ignored initial/terminal conditions ~~ leakage

DFT grid ~~ limited frequency resolution

improvements by windowing and interpolation

- the leakage is not eliminated
- the methods involve hyper-parameters

## Generalization of (SYS) to noisy data

preprocessing: rank-mL + n approx. of  $\mathcal{H}_L(w_d)$ 

- hyper-parameters  $L \ge \ell + 1$  and *n*
- if the approximation preserves the Hankel structure, the method is maximum-likelihood in the EIV setting

#### regularization with $||g||_1$

hyper-parameter: the 1-norm regularization parameter

regularization with the nuclear norm of  $\mathscr{H}_{L}(\widehat{w_{d}})$ 

hyper-parameters: L and the regularization parameter

## Matlab implementation

function Hh = dd\_frest(ud, yd, z, n)
L = n + 1; t = (1:L)';
m = size(ud, 2); p = size(yd, 2);

%% preprocessing by low-rank approximation
H = [moshank(ud, L); moshank(yd, L)];
[U, ~, ~] = svd(H); P = U(:, 1:m \* L + n);

%% form and solve the system of equations
for k = 1:length(z)
A = [[zeros(m\*L, p); -kron(z(k).^t, eye(p))] P];
hg = A \ [kron(z(k).^t, eye(m)); zeros(p\*L, m)];
Hh(:, :, k) = hg(1:p, :);
end

- effectively 5 lines of code
- MIMO case, multiple evaluation points
- L = n + 1 in order to have a single hyper-parameter

## Example: EIV setup with 4th order system

#### dd\_frest is compared with

- ident parametric maximum-likelihood estimator
- spa nonparameteric estimator with Welch filter



# Monte-Carlo simulation over different noise levels and number of samples



 $e_a := 100\% \cdot |(|\overline{H}_z| - |\widehat{H}_z|)| / |\overline{H}_z|$ 

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## Kernel representation LTI systems

$$\mathscr{B} = \ker R(\sigma) := \left\{ w \mid R(\sigma)w = 0 \right\}$$
$$= \left\{ w \mid R_0w + R_1\sigma w + \dots + R_\ell \sigma^\ell w = 0 \right\}$$

#### nonlinear time-invariant system

$$\mathscr{B} = \left\{ w \mid R(\underbrace{w, \sigma w, \ldots, \sigma^{\ell} w}_{x}) = 0 \right\}$$

linearly parameterized R

$$R(x) = \sum \theta_i \phi_i(x) = \theta^\top \phi(x), \quad \begin{array}{cc} \phi & -- & \text{model structure} \\ \theta & -- & \text{parameter vector} \end{array}$$

## Polynomial SISO NARX system

$$\mathscr{B}(\boldsymbol{\theta}) = \left\{ \boldsymbol{w} = \begin{bmatrix} \boldsymbol{u} \\ \boldsymbol{y} \end{bmatrix} \mid \boldsymbol{y} = f(\boldsymbol{u}, \boldsymbol{\sigma} \boldsymbol{w}, \dots, \boldsymbol{\sigma}^{\ell} \boldsymbol{w}) \right\}$$

split f into 1st order (linear) and other (nonlinear) terms

$$f(x) = heta_{\mathsf{li}}^{ op} x + heta_{\mathsf{nl}}^{ op} \phi_{\mathsf{nl}}(x)$$

 $\phi_{nl}$  — vector of monomials

## Special cases

## Hammerstein $\phi_{nl}(x) = \begin{bmatrix} \phi_u(u) & \phi_u(\sigma u) & \cdots & \phi_u(\sigma^\ell u) \end{bmatrix}^\top$

#### FIR Volterra

$$\phi_{\mathsf{nl}}(x) = \phi_{\mathsf{nl}}(x_u), \text{ where } x_u := \mathsf{vec}(u, \sigma u, \dots, \sigma^\ell u).$$

#### bilinear

$$\phi_{\mathsf{nl}}(x) = x_u \otimes x_y, \quad ext{where } x_y := \mathsf{vec}(y, \sigma y, \dots, \sigma^{\ell-1}y)$$

generalized bilinear

$$\phi_{\mathsf{nl}}(x) = \phi_{u,\mathsf{nl}}(x_u) \otimes x_y$$

## LTI embedding of polynomial NARX system

$$\mathscr{B}_{\mathsf{ext}}(\theta) := \left\{ \mathsf{w}_{\mathsf{ext}} = \begin{bmatrix} u \\ u_{\mathsf{nl}} \\ y \end{bmatrix} \mid \sigma^{\ell} \mathsf{y} = \theta_{\mathsf{li}}^{\top} \mathsf{x} + \theta_{\mathsf{nl}}^{\top} u_{\mathsf{nl}} \right\}$$

define:  $\Pi_w w_{ext} := w$  and  $\Pi_{u_{nl}} w_{ext} := u_{nl}$ fact:  $\mathscr{B}(\theta) \subseteq \Pi_w \mathscr{B}_{ext}(\theta)$ , moreover

 $\mathscr{B}(\theta) = \Pi_{w} \big\{ w_{\mathsf{ext}} \in \mathscr{B}_{\mathsf{ext}}(\theta) \mid \Pi_{u_{\mathsf{nl}}} w_{\mathsf{ext}} = \phi_{\mathsf{nl}}(x) \big\}$ 

# FIR Volterra data-driven simulation given

data  $w_d = (u_d, y_d)$  of lag- $\ell$  FIR Volterra system  $\mathscr{B}$  $\phi_{nl}$  — system's model structure

assume ID conditions for  $\mathscr{B}_{ext}$  hold

then,  $\mathscr{B}|_L = \operatorname{image} M$ , where

$$M(w_{\text{ini}}, u) := \mathscr{H}_{L}(\sigma^{\ell} y_{d}) \underbrace{ \begin{bmatrix} \mathscr{H}_{\ell}(w_{d}) \\ \mathscr{H}_{L}(\sigma^{\ell} u_{d}) \\ \mathscr{H}_{\ell}(\phi_{\text{nl}}(x_{u_{d}})) \\ \mathscr{H}_{L}(\sigma^{\ell} \phi_{\text{nl}}(x_{u_{d}})) \end{bmatrix}^{\dagger} \begin{bmatrix} w_{\text{ini}} \\ u \\ \varphi_{\text{nl}}(x_{u_{\text{ini}}}) \\ \varphi_{\text{nl}}(x_{u}) \end{bmatrix}}_{g}$$

proof

$$\begin{bmatrix} \mathscr{H}_{\ell}(w_{d}) \\ \mathscr{H}_{L}(\sigma^{\ell}u_{d}) \\ & \mathscr{H}_{\ell}(\phi_{nl}(x_{u_{d}})) \\ & \mathscr{H}_{L}(\sigma^{\ell}\phi_{nl}(x_{u_{d}})) \\ & & \mathscr{H}_{L}(\sigma^{\ell}y_{d}) \end{bmatrix} g = \begin{bmatrix} w_{ini} \\ u \\ & u \\ & \phi_{nl}(x_{u_{ini}}) \\ & \phi_{nl}(x_{u_{ini}}) \\ & \phi_{nl}(x_{u}) \\ & y \end{bmatrix} B3$$

- B1 constraint on g, such that  $w_{ini} \land (u, \mathscr{H}_L(\sigma^{\ell} y_d)g) \in \mathscr{B}_{ext}$ B2 constraint  $u_{nl} = \phi_{nl}(x) \iff \mathscr{B}_{ext} = \mathscr{B}(\theta)$
- B3 defines the to-be-computed output y

#### generalized bilinear models

also tractable because B2:  $u_{nl} = \phi_{nl}(x)$  is still linear in y