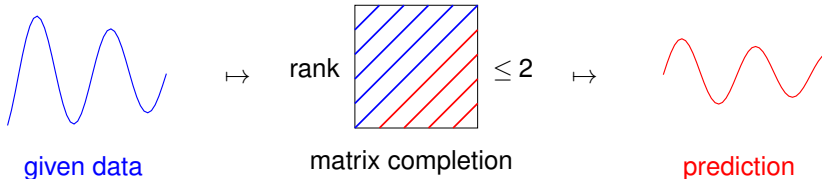


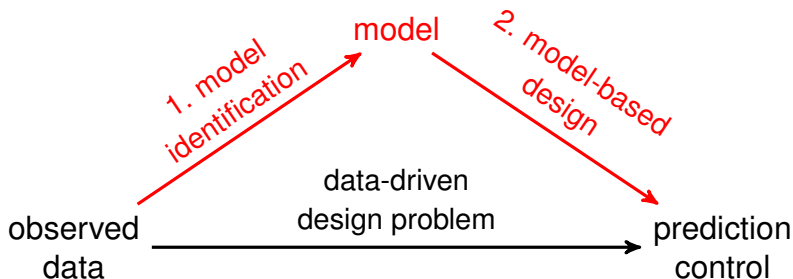
A missing data approach to data-driven filtering and control

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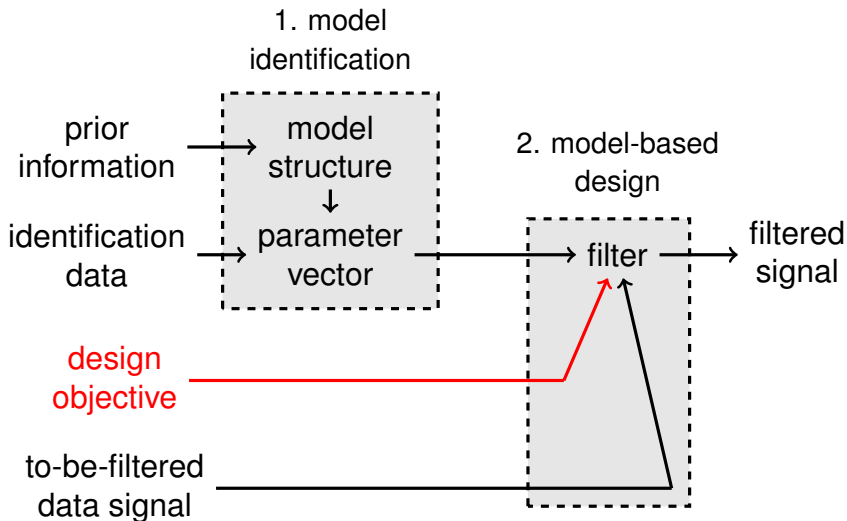
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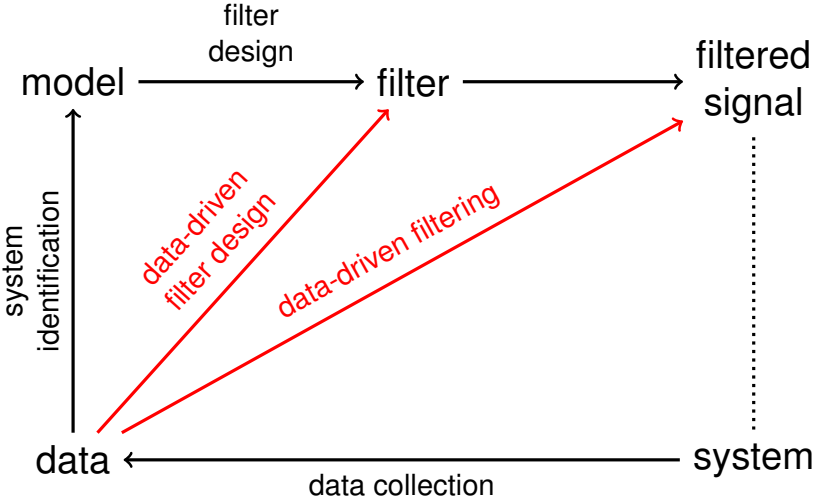
Modern filtering/control is model-based:
the design problem is split into two steps



System identification does not take into account the design objective



Data-driven methods avoid modeling



Combined modeling+design has benefits

identification ignores the design objective

the two-step approach is suboptimal

objective: define and solve a direct problem

observed data + filtering objective \mapsto filtered signal

Plan

Example: data-driven Kalman smoothing

Generalization: missing data estimation

Solution approach: matrix completion

Plan

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A dynamical system \mathcal{B} is a set of signals w

$$w \in \mathcal{B} \iff$$

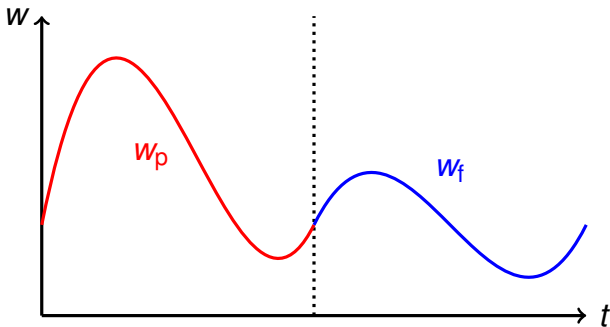
- ▶ the signal w is trajectory of the system \mathcal{B}
- ▶ \mathcal{B} is an exact model for w
- ▶ \mathcal{B} is unfalsified by w

we consider linear time-invariant systems $(w = \begin{bmatrix} u \\ y \end{bmatrix})$

\mathcal{L} — linear time-invariant model class

Initial conditions are specified by "past" traj.

$$W = W_p \wedge W_f$$



Representation free definition of smoothing

observer: given model \mathcal{B} and exact trajectory w_f

find w_p , such that $w_p \wedge w_f \in \mathcal{B}$

smoother: given model \mathcal{B} and noisy trajectory w_f

minimize $\|w_f - \hat{w}_f\|$ subject to $\hat{w}_p \wedge \hat{w}_f \in \mathcal{B}$ (MBS)

When does a trajectory $w_d \in \mathcal{B}$ specify \mathcal{B} ?

identifiability conditions

1. u_d is persistently exciting of "sufficiently high order"
2. \mathcal{B} is controllable

how to obtain \mathcal{B} back from w_d ?

$w_d \mapsto \mathcal{B}$ by choosing the simplest exact model for w_d

The most powerful unfalsified model of w_d , $\mathcal{B}_{\text{mpum}}(w_d)$ is the data generating system

complexity \leftrightarrow # inputs m and # states n

$$c(\mathcal{B}) = (m, n)$$

the most powerful unfalsified model

$$\mathcal{B}_{\text{mpum}}(w_d) := \arg \min_{\underbrace{\hat{\mathcal{B}} \in \mathcal{L}}_{\text{most powerful}}} c(\hat{\mathcal{B}}) \quad \text{subject to} \quad \underbrace{w_d \in \hat{\mathcal{B}}}_{\text{unfalsified model}}$$

$\mathcal{L}_{m,n}$ — set of models with complexity bounded by (m, n)

Data-driven smoothing replaces the model \mathcal{B} by trajectory $w_d \in \mathcal{B}$

observer: given trajectories w_d and w_f of \mathcal{B}

find w_p , such that $w_p \wedge w_f \in \mathcal{B}_{\text{mpum}}(w_d)$

smoother: given noisy traj. w_d and w_f of \mathcal{B} and (m, ℓ)

minimize $\underbrace{\|w_f - \hat{w}_f\|_2^2}_{\text{estimation error}} + \underbrace{\|w_d - \hat{w}_d\|_2^2}_{\text{identification error}} \quad (\text{DDS})$

subject to $\hat{w}_p \wedge \hat{w}_f \in \mathcal{B}_{\text{mpum}}(\hat{w}_d) \in \mathcal{L}_{m, \ell}$

Classical approach: divide and conquer

1. **identification**: given w_d and (m, ℓ)

$$\text{minimize } \|w_d - \hat{w}_d\| \quad \text{subject to } \mathcal{B}_{\text{mpum}}(\hat{w}_d) \in \mathcal{L}_{m,\ell}$$

2. **model-based filtering**: given w_f and $\hat{\mathcal{B}} := \mathcal{B}_{\text{mpum}}(\hat{w}_d)$

$$\text{minimize } \|w_f - \hat{w}_f\| \quad \text{subject to } \hat{w}_p \wedge \hat{w}_f \in \hat{\mathcal{B}}$$

Summary

model-based smoothing

given model \mathcal{B} and trajectory w_f

$$\text{minimize } \|w_f - \hat{w}_f\| \quad \text{subject to } \hat{w}_p \wedge \hat{w}_f \in \mathcal{B} \quad (\text{MBS})$$

data-driven smoothing

given trajectories w_d and w_f and complexity (m, ℓ)

$$\begin{aligned} &\text{minimize } \|w_f - \hat{w}_f\|_2^2 + \|w_d - \hat{w}_d\|_2^2 \\ &\text{subject to } \hat{w}_p \wedge \hat{w}_f \in \mathcal{B}_{\text{mpum}}(\hat{w}_d) \in \mathcal{L}_{m,\ell} \end{aligned} \quad (\text{DDS})$$

Plan

Example: data-driven Kalman smoothing

Generalization: missing data estimation

Solution approach: matrix completion

We aim to find missing part of trajectory

missing data — interpolated from $w \in \mathcal{B}$

exact data— kept fixed

inexact / "noisy" data — approximated by $\min \|\text{error}\|_2$

Other examples fit in the same setting

? — missing, E — exact, N — noisy
 $w = \Pi \begin{bmatrix} u \\ y \end{bmatrix}$, u — input, y — output

example	w_p	u_f	y_f
state estimation	?	E	E
EIV Kalman smoothing	?	N	N
classical Kalman smoothing	?	E	N
simulation	E	E	?
partial realization	E	E	E/?
noisy realization	E	E	N/?
output tracking	E	?	N

classical Kalman filter

	past	future
input	?	u
output	?	y

output tracking control

	past	future
input	u_p	?
output	y_p	y_{ref}

$$\begin{aligned} & \text{minimize} && \|y - \hat{y}\| \\ & \text{subject to} && w_p \wedge (u, \hat{y}) \in \mathcal{B} \end{aligned}$$

$$\begin{aligned} & \text{minimize} && \underbrace{\|y_{ref} - \hat{y}\|}_{\text{tracking error}} \\ & \text{subject to} && w_p \wedge (\hat{u}, \hat{y}) \in \mathcal{B} \end{aligned}$$

Weighted approximation criterion accounts for exact, missing, and noisy data

error vector: $\mathbf{e} := \mathbf{w} - \hat{\mathbf{w}}$

$$\|\mathbf{e}\|_v := \sqrt{\sum_t \sum_i v_i(t) e_i^2(t)}$$

weight	used for	to	by
$v_i(t) = \infty$	$w_i(t)$ exact	interpolate $w_i(t)$	$\mathbf{e}_i(t) = 0$
$v_i(t) \in (0, \infty)$	$w_i(t)$ noisy	approx. $w_i(t)$	$\min \ \mathbf{e}_i(t)\ $
$v_i(t) = 0$	$w_i(t)$ missing	fill in $w_i(t)$	$\hat{\mathbf{w}} \in \hat{\mathcal{B}}$

Data-driven signal processing can be posed as missing data estimation problem

$$\begin{array}{ll} \text{minimize} & \|w_d - \hat{w}_d\|_2^2 + \|w - \hat{w}\|_v^2 \\ \text{subject to} & \hat{w} \in \mathcal{B}_{\text{mpum}}(\hat{w}_d) \in \mathcal{L}_{m,l} \end{array} \quad (\text{DD-SP})$$

the recovered missing values of \hat{w} are the desired result

Plan

Example: data-driven Kalman smoothing

Generalization: missing data estimation

Solution approach: matrix completion

$w \in \mathcal{B} \iff$ Hankel matrix is low-rank

exact trajectory $w \in \mathcal{B} \in \mathcal{L}_{m,\ell}$

\iff

$$R_0 w(t) + R_1 w(t+1) + \dots + R_\ell w(t+\ell) = 0$$

\iff

rank deficient

$$\mathcal{H}(w) := \begin{bmatrix} w(1) & w(2) & \dots & w(T-\ell) \\ w(2) & w(3) & \dots & w(T-\ell+1) \\ w(3) & w(4) & \dots & w(T-\ell+2) \\ \vdots & \vdots & & \vdots \\ w(\ell+1) & w(\ell+2) & \dots & w(T) \end{bmatrix}$$

relation at time $t = 1$

$$R_0 w(1) + R_1 w(2) + \dots + R_\ell w(\ell + 1) = 0$$

in matrix form:

$$\begin{bmatrix} R_0 & R_1 & \dots & R_\ell \end{bmatrix} \begin{bmatrix} w(1) \\ w(2) \\ \vdots \\ w(\ell + 1) \end{bmatrix} = 0$$

relation at time $t = 2$

$$R_0 w(2) + R_1 w(3) + \cdots + R_\ell w(\ell + 2) = 0$$

in matrix form:

$$\begin{bmatrix} R_0 & R_1 & \cdots & R_\ell \end{bmatrix} \begin{bmatrix} w(2) \\ w(3) \\ \vdots \\ w(\ell + 2) \end{bmatrix} = 0$$

relation at time $t = T - \ell$

$$R_0 w(T - \ell) + R_1 w(T - \ell + 1) + \dots + R_\ell w(T) = 0$$

in matrix form:

$$\begin{bmatrix} R_0 & R_1 & \dots & R_\ell \end{bmatrix} \begin{bmatrix} w(T - \ell) \\ w(T - \ell + 1) \\ w(T - \ell + 2) \\ \vdots \\ w(T) \end{bmatrix} = 0$$

relation for $t = 1, \dots, T - \ell$

$$R_0 w(t) + R_1 w(t+1) + \dots + R_\ell w(t+\ell) = 0$$

in matrix form:

$$\underbrace{\begin{bmatrix} R_0 & R_1 & \dots & R_\ell \end{bmatrix}}_R \underbrace{\begin{bmatrix} w(1) & w(2) & \dots & w(T-\ell) \\ w(2) & w(3) & \dots & w(T-\ell+1) \\ w(3) & w(4) & \dots & w(T-\ell+2) \\ \vdots & \vdots & & \vdots \\ w(\ell+1) & w(\ell+2) & \dots & w(T) \end{bmatrix}}_{\mathcal{H}(w)} = 0$$

$$w \in \mathcal{B} \in \mathcal{L}_{m,l}$$



there is $R \in \mathbb{R}^{(q-m) \times q(\ell+1)}$ full row rank,

$$\text{such that } R\mathcal{H}(w) = 0$$



$$\text{rank}(\mathcal{H}(w)) \leq q\ell + m$$

q — # of variables

$\widehat{\mathbf{W}} \in \mathcal{B}_{\text{mpum}}(\widehat{\mathbf{W}}_d)$ is equivalent to rank constraint on a mosaic-Hankel matrix

$$\begin{aligned} \widehat{\mathbf{W}} \in \mathcal{B}_{\text{mpum}}(\widehat{\mathbf{W}}_d) \in \mathcal{L}_{m,l} \\ \Downarrow \\ \widehat{\mathbf{W}}_d \in \widehat{\mathcal{B}} \in \mathcal{L}_{m,l} \quad \text{and} \quad \widehat{\mathbf{W}} \in \widehat{\mathcal{B}} \\ \Updownarrow \\ \text{rank} \left(\underbrace{\begin{bmatrix} \mathcal{H}(\widehat{\mathbf{W}}_d) & \mathcal{H}(\widehat{\mathbf{W}}) \end{bmatrix}}_{\mathcal{H}(\widehat{\mathbf{W}}_d, \widehat{\mathbf{W}})} \right) \leq ql + m \end{aligned}$$

Data-driven signal processing

\iff structured low-rank approximation

$$\begin{aligned} &\text{minimize} && \|w_d - \hat{w}_d\|_2^2 + \|w - \hat{w}\|_v^2 \\ &\text{subject to} && \hat{w} \in \mathcal{B}_{\text{mpum}}(\hat{w}_d) \in \mathcal{L}_{m,l} \end{aligned}$$

\iff

$$\begin{aligned} &\text{minimize} && \|w' - \hat{w}'\|_{v'} \\ &\text{subject to} && \text{rank}(\mathcal{H}(\hat{w}')) \leq r \end{aligned}$$

Three main classes of solution methods

local optimization

nuclear norm relaxation

subspace methods

considerations

- ▶ generality
- ▶ user defined hyper parameters
- ▶ availability of efficient algorithms/software

Local optimization using variable projections: analytical elimination of \hat{w}

kernel representation

$$\min_{R \text{ f.r.r.}} \left(\min_{\hat{w}} \|w - \hat{w}\| \text{ subject to } R\mathcal{H}(\hat{w}) = 0 \right)$$

variable projection (VARPRO): elimination of \hat{w} leads to

$$\text{minimize } f(R) \text{ subject to } R \text{ full row rank}$$

Dealing with the " R full row rank" constraint

1. impose a quadratic equality constraint $RR^T = I$
2. using specialized methods for optimization on a manifold
3. R full row rank $\iff R\Pi = \begin{bmatrix} X & I \end{bmatrix}$ with Π permutation
 - ▶ Π fixed \rightsquigarrow total least-squares
 - ▶ Π can be changed during the optimization

Summary of the VARPRO approach

kernel representation \rightsquigarrow parameter opt. problem

$$\min_{\hat{w}, R \text{ f.r.r.}} \|w - \hat{w}\| \quad \text{subject to} \quad R\mathcal{H}(\hat{w}) = 0$$

elimination of \hat{w} \rightsquigarrow optimization on a manifold

$$\min_{R \text{ f.r.r.}} f(R)$$

in case of mosaic-Hankel \mathcal{H} , f can be evaluated fast

Numerical example with Kalman smoothing

simulation setup

- ▶ $\overline{\mathcal{B}} \in \mathcal{L}_{1,2}$ — 2nd order LTI system
- ▶ $w_f = \overline{w}_f + \text{noise}$, $\overline{w}_f \in \mathcal{B}$ — step response
- ▶ $w_d = \overline{w}_d + \text{noise}$, $\overline{w}_d \in \mathcal{B}$

smoothing with known model

- ▶ state space solution
- ▶ solution of (MBS)

smoothing with unknown model

- ▶ identification + model-based design
- ▶ solution of (DDS)

Known model: the missing data approach (MBS) recovers the state space solution

state space solution

$$\text{minimize} \quad \left\| \begin{bmatrix} u_f \\ y_f \end{bmatrix} - \begin{bmatrix} 0 & I \\ \mathcal{O}_T(A, C) & \mathcal{F}_T(H) \end{bmatrix} \begin{bmatrix} \hat{x}_{\text{ini}} \\ \hat{u}_f \end{bmatrix} \right\| \quad (\text{SSS})$$

representation free solution

(MBS) is a generalized least squares

approximation error $e := (\|\bar{w}_f - \hat{w}_f\|) / \|\bar{w}_f\|$

method	(MBS)	(SSS)
error e	0.083653	0.083653

Unknown model: (DDS) gives better results than the model-based approach

classical approach

identification + (SSS)

data-driven approach

solution of (DDS) with local optimization

simulation result

method	(MBS)	(DDS)	classical
error e	0.083653	0.087705	0.091948

Conclusion

motivation: combine the modeling and design problems

we aim to find the missing part of a trajectory $w \in \mathcal{B}$

reformulation as weighted structured low-rank approx.

Future work

statistical analysis

computational efficiency / recursive computation

other methods: subspace, convex relaxation, . . .